

GR 98 P 2958

1/PRTS

## Description

## Method and arrangement for determining parameters of a technical system

5           The invention relates to a method and an arrangement for determining parameters of a technical system.

          During multichannel transmission and multichannel reception of signals, interference  
10 frequently occurs, for example, between the signals/images. One typical example in this case is mixing of a voice signal with noise, which can present a major problem in telecommunications and in video conferences. The present invention thus relates to the  
15 field of signal separation in order, for example, to recover an original voice signal.

          Typical known techniques for separation of source signals based on mixed signals are based on time averaging or filtering of the signals. This  
20 intrinsically has disadvantages in terms of the computation complexity.

          Methods based on so-called blind channel equalization (signal equalization without prior knowledge of the transmission channel) are also known,  
25 but these methods always require a certain amount of knowledge about the source signals, such as knowledge about their statistical distribution.

          The problem of signal separation also occurs, for example, when two speakers are speaking into two  
30 microphones positioned at a distance from one another, so that each microphone receives a mixture of the signals spoken by the two speakers. The problem thus arises of separating the signal mixture once again, that is to say of separating a set of superimposed  
35 input signals. L. Molgedey, H.G. Schuster, "Separation of a Mixture of Independent Signals using Time-Delayed Correlations", Phys. Ref. Lett. 72, 3634 (1994) in this case discloses the following method: the problem of

0903043-042701

separating  $n$  superimposed and correlated source signals (input signals) and at the same time of establishing mixing coefficients of the source intensities can be reduced to an intrinsic value problem, in which two  
5 symmetrical  $n \times n$  matrices must be diagonalized simultaneously. The matrix elements are measurable time-delayed correlation functions. This intrinsic value problem can be solved by means of a neural network, in which case the learning rules for the  
10 lateral inhibiting interactions between the neurons can be established by means of a Liapunov function whose minima provide the (degenerate) solutions to the problem.

This method has also already been applied to  
15 the acoustic input signals (see F. Ehlers, H.G. Schuster, "Blind Separation of convolutive mixtures and an application in automatic speech recognition", IEEE Trans. Signal Proc. (1997)).

DE 195 31 388 C1 discloses a signal separation  
20 method and a signal separation device for nonlinear mixtures of unknown signals (blind channel), which is illustrated schematically in Figure 3.

This German Patent relates to the separation of a signal mixture comprising the nonlinear  
25 superimposition of  $M$  unknown source signals  $X_1, X_2$ , where  $N$  ( $N \geq M$ ) different mixtures of  $M$  source signals  $X_1, X_2$  including any interference signal which may be present are supplied to a signal evaluation device, which analyzes the signal statistically to establish  
30 the nonlinear transmission factors and using these calculated factors to reverse the mixing process, so that the  $N$  outputs of the signal separation device contain, as approximately as possible, the  $M$  source signals without superimpositions.

093041-042701  
FDZ240-ET40E660

It thus becomes possible to treat nonlinear mixtures, in which the case the term nonlinear means that the source signals  $X_1$ ,  $X_2$  are mixed by means of an unknown nonlinear system  $G$ . The unknown system  $G$  is described  
5 by a so-called Volterra series, and the signal separation device  $G-1$  establishes the coefficients in the Volterra series. Once this

09830413 042701  
FOI2240 EF40E860

is known, it is possible to unmix the signal mixture. Furthermore, the coefficients can be used for further analysis in order to determine the position or speed of the signal sources.

5           The method which is known from this document in this case essentially comprises two steps:

- Firstly, the nonlinear equations which are selected uniquely by the selectable degree of nonlinearity in the mixing process are solved by a sliding time window, and the solutions are averaged over this time. This time averaging process represents a major disadvantage of this known technique, since it increases the computation complexity, while at the same time increasing the time required for the calculation process.
- Secondly, the potential formed from a sufficiently large number of different cumulants of the estimated output signals is minimized, with the values required to calculate the potential originating from a sliding time window whose length can be varied. In this case, it is assumed that the mixing system varies sufficiently slowly that this change can be ignored in the calculation of the sought mixing factors. According to this German Patent, the second said step is carried out by constructing and minimizing a cost function. When the global minimum is reached, the optimum values, in this case the transmission factors, have been found.

30           With regard to the time involved and the computation complexity, the method described in DE 195 31 388 C1 is disadvantageous, since the time averaging process has to be carried out at the end of the first method step mentioned above.

35           The present invention is thus based on the object of providing a method and an arrangement which allow the separation of superimposed, statistically mutually independent, acoustic signals with reduced computation complexity.

09830413, 042701

This object is achieved by the method and by the arrangement having the features according to the independent claims.

5 In a method for determining parameters of a technical system, by means of which method output signals can be determined from a set of superimposed, statistically mutually independent input signals, the parameters are determined in such a manner that the statistical independence of the output signals is  
10 maximized.

An arrangement for determining parameters of a technical system, by means of which system output signals can be determined from a set of superimposed, statistically mutually independent input signals, has a  
15 processor which is set up in such a manner that the parameters can be determined in such a manner that the statistical independence of the output signals is maximized.

Advantageous developments of the invention  
20 result from the dependent claims.

The parameters are preferably determined using an iterative method.

In a further refinement, the parameters are elements in an unmixing matrix, by which the set of  
25 superimposed input signals is multiplied or else convoluted, by which means the output signals are formed.

The optimization of the parameters in the unmixing matrix is preferably obtained by the following  
30 steps:

- repetition of a time-delayed decorrelation calculation in order to determine the intrinsic values in the unmixing matrix,

09330413.042701

- determination of the intrinsic values in the unmixing matrix for which cross-correlations assume a minimum value, and

- carrying out cumulant minimization, with the  
5 intrinsic values determined in the previous step being used as start values for the cumulant minimization.

The cumulant minimization can be used, for example, by training a neural network, or else by any other known minimization technique, such as gradient  
10 descent or Monte Carlo simulations.

In one development, at least one diagonal parameter of the unmixing matrix is set to a predetermined value during the optimization of the parameters in the unmixing matrix, thus ensuring the  
15 stability of the minimization process with respect to a global minimum.

The unmixing matrix is preferably limited to a finite impulse response, that is to say an FIR filter (Finite Impulse Response) is used to form the  
20 individual components of the unmixing matrix. The FIR filter may be either a causal FIR filter or else a non-causal FIR filter.

Furthermore, the unmixing matrix is preferably stabilized by projection on to a unit circle during the  
25 cumulant minimization process.

The developments apply not only to the method but also to the arrangement, in which case, in the case of the arrangement, the processor is in each case set up in such a manner that the corresponding method step  
30 can be or is carried out.

The invention and its developments can advantageously be used for separation of superimposed, statistically mutually independent input signals, in particular acoustic input signals.

The method and the arrangement can be used for any desired number of input signals.

Further advantages, features and characteristics of the present invention will now be explained in more detail using an exemplary embodiment and with reference to the attached figures of the drawings, in which:

Figure 1 shows the use of a system for separation of superimposed, statistically mutually independent acoustic signals according to the exemplary embodiment,

Figure 2 shows a schematic illustration of the system from Figure 1, and

Figure 3 shows a signal separation device, which is known from the prior art (DE 195 31 388 C1), for nonlinear mixtures of unknown signals.

The statistical independence between the source signals (the original voice signal and the noise), also referred to as input signals in the following text, is used to recover the original voice signal from a mixture of signals, and the inverse process to that of the dynamic system, which has resulted in the mixing of the signals, is trained essentially approximately (is learnt). Two different mixtures of the voice signal and of the noise signal are obtained, for example, by means of two microphones 1, 2 (see Figure 1) which are at a distance from one another and/or are aligned in opposite directions. The so-called time-delayed decorrelation technique (TDD) is used to initiate the learning phase in the method, that is to say in order to determine and specify start values for the learning phase, which allows the computation complexity for cumulant minimization as described in the following text to be reduced, and allows the risk of local minima to be reduced.

09830413 042701

Figure 1 shows two microphones 1, 2, which pick up a first input signal  $Z1(t)$  and a second input signal  $Z2(t)$ . These input signals  $Z1(t)$  and  $Z2(t)$  can in turn each be mixed with one another and with noise, as is represented symbolically by a mixing matrix  $S$  (see reference symbol 3) in Figure 1. After reception and/or transmission, a set  $X1(t)$  and  $X2(t)$  of superimposed, statistically mutually independent input signals  $Z1(t)$  and  $Z2(t)$  is obtained. These signals are entered in a calculation unit 4, in which essentially two steps are carried out, as is represented symbolically by a calculation unit B (reference symbol 6) for the first step and a neural network 5 for the second step.

The calculation unit 4 determines two output signals  $Y1(t)$  and  $Y2(t)$ , respectively, which are approximately equal to the input signals  $Z1(t)$  and  $Z2(t)$ , respectively, when the parameters are set optimally in the calculation unit 4. In other words, when the parameters of the matrices which are used are set optimally in the calculation unit 4, this calculation unit 4 essentially carries out the inverse process to that of the dynamic mixing process, which is represented symbolically by the matrix  $S$  (reference symbol 3). The exemplary embodiment relates to the optimization process for setting the parameters for the unmixing matrix.

The parameters of the matrices in the calculation unit 4 are in this case optimized such that the statistical independence between the output signals  $Y1(t)$ ,  $Y2(t)$  obtained by the matrix process in the calculation unit 4 is maximized. To this end, the output signals  $Y1(t)$  and  $Y2(t)$ , respectively, are fed back to the calculation unit 4 (see the feedback loops 7 and 8, respectively). An iterative method is used to determine whether the statistical independence of the output signal  $Y1(t)$  and  $Y2(t)$ ,

09330413 "042701



respectively, has increased in comparison to the previous iteration step (so that the iteration is in the "right" direction, in the direction of the global minimum of a cost function, which will be described in the following text), or not.

5

09830413.042701

Figure 2 shows a mathematical representation of the layout from Figure 1, in which case the mixing process 3 can be described mathematically by a matrix  $S(q)$ , and the unmixing process, which is intended to be carried out by the calculation unit 4, is symbolized by an unmixing matrix  $M(q)$ .

Figure 2 thus illustrates the problem of separation of a so-called multichannel blind source (multiple channel source without a-priori knowledge) into two dimensions. In this case, it is assumed that the mixing system  $S(q)$ , where  $q$  represents a unit delay, is stable and, at the same time, also has a stable inversion, that is to say it is a minimal phase system. Furthermore, it is assumed that the input signals  $Z_1(t)$  and  $Z_2(t)$  (for example a voice signal and a noise signal) are statistically mutually independent and do not have a Gaussian distribution. The sets  $X_1(t)$  and  $X_2(t)$  of superimposed input signals  $Z_1(t)$  and  $Z_2(t)$  are input signals to an unmixing system having an unmixing matrix  $M(q)$  whose parameters (matrix elements) are trained to maximize the statistical independence between the output signals  $Y_1(t)$  and  $Y_2(t)$ . In this case, the term "training" means the well known learning process of, for example, a neuron network, which should be cited as an example of a technique to maximize the statistical independence. This is done by minimizing a cost function  $J(M)$ , which will be described further below.

A cumulant cost function is formed, which minimizes the diagonal cumulant elements of the cumulant order 2-4:

$$D_{\text{cum}} \approx J(M) = \sum_{i=1}^4 \sum_{\text{nondiag}} [c^{(i)}_{\text{nondiag}}]^p$$

The following aspects of dynamic mixing by the mixing matrix  $S(q)$  need to be taken into account in this case:

- Stability of the unmixing system:

This is achieved by limiting  $M(q)$  to a finite impulse response (FIR filter). The stability of the FIR system  $M(q)$  can, furthermore, also be obtained by carrying out a projection on to a unit circle during the learning phase. Any non-causality of the inversion of  $S(q)$  which may be present can be compensated for by a suitable time shift (delay) to the input signal  $X(t)$ .

- Uniqueness of the separated signals  $Y(t)$ :

In the case of steady-state mixing processes, the original source signals are recovered by scaling. For dynamic unmixing, the risk of the separated signals  $Y(t)$  not being unique is even greater. It is obvious that, in the situation where  $Y_1(t)$  and  $Y_2(t)$  are statistically mutually independent, any linear-filtered modification of these signals will also still be statistically independent. Additional information is therefore required in order to reduce the inherent ambiguity of the problem.

- Gaussian deformation of the data:

Algorithms on a cumulant basis for steady-state blind source separation effectively minimize or eliminate higher-order diagonal cumulants corresponding to the output signals  $Y(t)$ . On the other hand, linear filtering leads to the data being deformed to a Gaussian distribution, with the higher-order cumulants moving in the 0 direction. This can thus lead to limit solutions, in which the cost function reaches local minimum, but with the desired actual separation (global minimum) not being achieved. In order to avoid this undesirable situation, the structure of the unmixing transfer function (unmixing matrix)  $M(q)$  is subject to a number of limitations.

09330413 "042701

In order to avoid the abovementioned problems, an approach is chosen in which at least one (or else all) of the diagonal elements is or are set to the unit value:

- 5  $M_{11}(q) = 1$   
and  
 $M_{22}(q) = 1.$

This assumption is exact if the mixing elements  $S_{11}(q)$  and/or  $S_{22}(q)$  likewise have a unit value "1".  
10 Otherwise, it is assumed that  $S_{11}(q)$  and/or  $S_{22}(q)$  have stable inversions, which allows the diagonal elements to be scaled from  $M(q)$  to the unit value. This approach considerably reduces the ambiguity of a solution and, furthermore, effectively avoids the risk of excessive  
15 Gaussian-distribution deformation of the output signals. Even if, as stated above, the limitation to the diagonal elements of  $M(q)$  is at first glance to be highly restrictive, this assumption is generally satisfied in practical use. One typical example is the  
20 removal of noise from voice signals on the basis of a recording using two microphones, with the microphones being physically separated from one another or one microphone pointing in the direction of the speaker, while the other microphone points in the opposite  
25 direction, so that the second signal, facing away from the speaker, essentially includes only a noise signal.

The cumulant approach is based on the direct determination of the diagonal cumulant, as is stated in the article mentioned initially by F. Ehlers and  
30 H. Schuster, "Blind Separation of convolutive mixtures and an application in automatic speech recognition", IEEE Trans. Signal Proc. (1997), although this inherently has the disadvantage that numerical solution is highly complex. Suitable initialization is thus used  
35 for this minimization method. In order to determine start values

09830413, 042701

for the minimization method, the present invention uses the technique of time-delay decorrelation (TDD) for simultaneous decorrelation of two different time delays, in which case this TDD technique can be based on a suitable matrix intrinsic-value problem. As already stated, this TDD technique is used according to the present invention for initiation of the diagonal (cross-correlation) cumulant minimization problem.

In summary, the method can be subdivided into the two following steps:

1. Repetition of the TDD method on the basis of the intrinsic-value problem in the frequency domain for different delay pairs and determination of that solution for which the cross-correlation terms have a minimum value.
2. Initiation (start) of the diagonal cumulant minimization process on the basis of the start values (FIR parameters) determined in the above step.

A number of major characteristics and advantages will be summarized once again in the following text:

- No a-priori knowledge of the signal characteristics is required, with the exception of the necessity for statistical independence.
- The stability of the dynamic unmixing system is ensured by the modulation of its components as an FIR filter.
- Excessive Gaussian distribution deformation is avoided by the approach of at least one of the elements in the mixing transfer function matrix (unmixing matrix) being set to the unit value, or being able to be scaled to the unit value, and
- since the cumulant minimization step (step 2) requires a large amount of computation complexity,

the learning algorithm, for example of a neural network, is initialized using the TDD method.

09830413.042701

The following text contains a program in Matlab, Version 4 or Version [lacuna], by means of which the exemplary embodiment described above can be implemented on a computer:

```

5  function [cost, out1, out2] = cumulant_costFIRa2 (par, input, p1, p11, p2, p
    22, a3, a4);
    % [cost, out1, out2] = cumulant_costFIRa2 (par, input, p1, p11, p2, p22, a3, a
    4);
10  % cumulant cost
    % FIR representation used
    % filter function used in both directions (non_causal)

    [np, mp] = size (par);
15  fir1 = par (1:p1);
    fir11 = par (1+p1:p1+p11);
    fir2 = par (1+p1+p11:p1+p11+p2);
    fir22 = par (p1+p11+p2+1:mp);
    den = 1;                                %FIR only
20  out1 = [input (:, 1) - filter (fir1, den, input (:, 2)) - flipud (filter ([0
    fir11], [den], flipud (input (:, 2))))];
                                     %/std (input (1, :));           %dlsim

25  %filter
    out2 = [input (:, 2) - filter (fir2, den, input (:, 1)) - flipud (filter ([0
    fir22], [den],
    flipud (input (:, 1))))]; %/std (input (:, 2));
                                     %dlsim      %filter

30  out [out1 out2];
    %out1 = out1/std (out1); % this scaling was not needed in examples
    in SIP98 paper
    %out2 = out2/std (out2);

35  Ld = 0; % number of delays in calculating the cross-correlation
    cost3 = 0;
    cost4 = 0;
    costALL1 = [];

```

10/24/01 EFT03350

1943-1944